

**Radio and Optical Wave Propagation – Prof. L. Luini,
July 1st, 2016**

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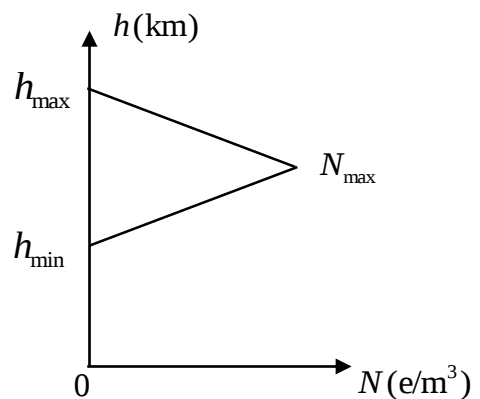
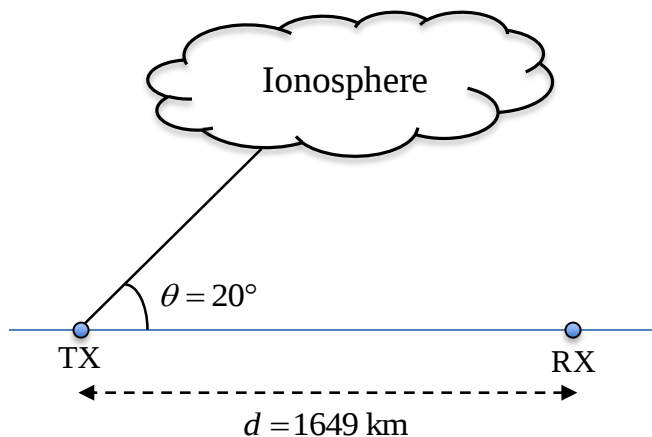
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Exercise 1

Making reference to the figure below, the transmitter TX, working at $f_1 = 52.6$ MHz, reaches the user RX, at a distance $d = 1649$ km, by exploiting the ionosphere. TX transmits with elevation angle $\theta = 20^\circ$. The ionosphere can be modelled with the symmetric electron density profile sketched in the figure (right side), where $h_{\max} = 400$ km and $h_{\min} = 100$ km.

- 1) Calculate the maximum electron density $N_{\max}$.
- 2) Considering the same ionospheric profile and that the frequency of the transmitter increases to $f_2 = 150$ MHz, evaluate if TX can reach a geostationary satellite seen at elevation angle of 10° .

Assume: the virtual reflection height h_V is 1.2 of the height at which the wave is actually reflected.



Solution:

Considering the figure below, given the distance between the TX and RX and the elevation angle, the virtual reflection height h_V is given by:

$$h_V = \frac{d}{2} \tan \theta = 300 \text{ km}$$

Actual reflection occurs at $h_R = h_V/1.2 = 250$ km. This is also the height at which $N_{\max}$ lies:

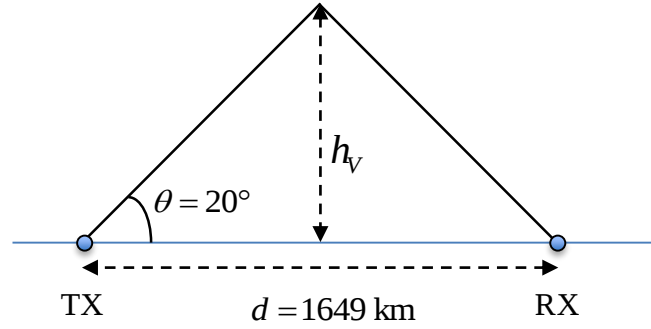
$$h_N = h_{\min} + (h_{\max} - h_{\min})/2 = 250 \text{ km}$$

The value of $N_{\max}$ can be derived from:

$$\cos \theta = \sqrt{1 - \left(\frac{f_p}{f_1} \right)^2} = \sqrt{1 - \left(\frac{9\sqrt{N_{\max}}}{f_1} \right)^2}$$

The inversion of such equation yields:

$$N_{\max} = \frac{(1 - (\cos \theta)^2) f_1^2}{81} \approx 4 \times 10^{12} \text{ e/m}^3$$



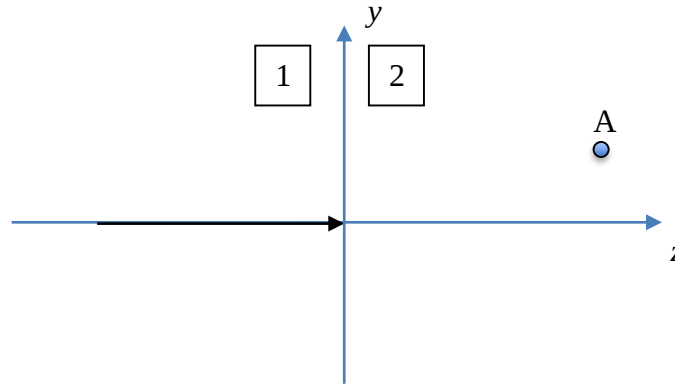
If the transmission frequency increases to $f_2 = 150$ MHz in the same ionospheric conditions, the maximum elevation angle to obtain total reflection decreases to:

$$\theta_{\max} = \cos^{-1} \left[\sqrt{1 - \left(\frac{f_p}{f_2} \right)^2} \right] = \cos^{-1} \left[\sqrt{1 - \left(\frac{9\sqrt{N_{\max}}}{f_2} \right)^2} \right] \approx 6.9^\circ$$

Being the new elevation angle of the link equal to 10° , i.e. higher than $\theta_{\max}$, the wave will not be reflected by the ionosphere (only partially refracted) and the TX will be able to reach the GEO satellite.

Exercise 2

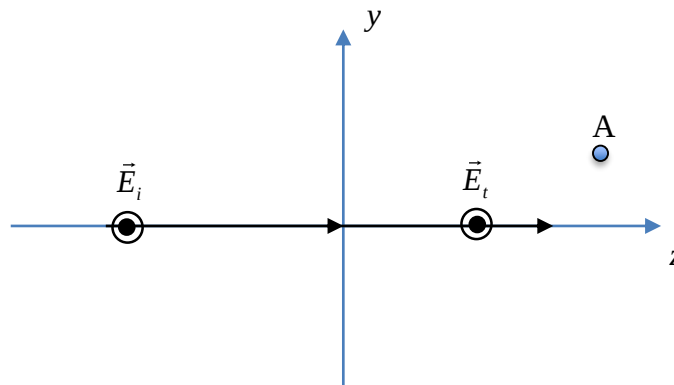
A plane sinusoidal wave at $f = 3$ GHz propagates in the vacuum and impinges orthogonally on a medium characterized by $\epsilon_{r2} = 4-j4$, $\mu_{r2} = 1$. The absolute value of the electric field in ($z = 0$ m) is $E_0 = 5$ V/m (assume phase equal to zero). Write the full expression of the magnetic field in the second medium and calculate the power received by the antenna in A($z = 0.2$ m, $y = 0.01$ m) that has equivalent area $A_E = 1$ m².



Solution:

The problem concerns a TEM plane wave: the choice of the wave polarization is irrelevant. Let's assume the electric field is as the one in the figure below, i.e.:

$$\vec{E}_i = -E_0 \vec{\mu}_x e^{-j\beta_1 z} \text{ V/m}$$



The intrinsic impedances of the two media (no approximations are possible for the lossy medium) are:

$$\eta_1 = \eta_0 = 377 \text{ } \Omega$$

$$\eta_2 = \sqrt{\frac{j\omega\mu_0\mu_{r2}}{j\omega\epsilon_0\epsilon_{r2}}} = 146.3 + j60.6 \text{ } \Omega$$

The reflection coefficient is:

$$\Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = -0.4271 + j0.1647$$

The electric field at the boundary between the two media is given by:

$$\vec{E}_t(0) = \vec{E}_i (1 + \Gamma) = -5\vec{\mu}_x (0.5783 + j0.1647) = (-2.8916 - j0.8233)\vec{\mu}_x \text{ V/m}$$

The full expression of the electric field in the second medium is:

$$\vec{E}_t(z) = \vec{E}_t(0)e^{-\gamma_2 z} = \vec{E}_t(0)e^{-(\alpha_2 + j\beta_2)z} \quad \text{V/m} \rightarrow \vec{E}_t(z = 0.2 \text{ m}) = \vec{E}_t(0)e^{-\gamma_2 0.2} = 1.95 \times 10^{-5} + j2.56 \times 10^{-5} \text{ V/m}$$

The propagation constant in the second medium is:

$$\gamma_2 = \sqrt{j\omega\mu j\omega\epsilon} = 57.2 + j138.2 \text{ m}^{-1}$$

Thus, the full expression of the magnetic field in the second medium is:

$$\vec{H}_t(z) = \frac{(-2.8916 - j0.8233)\vec{\mu}_y}{\eta_2} e^{-\gamma_2 z} = (-0.0189 + j0.0022)\vec{\mu}_y e^{-(\alpha_2 + j\beta_2)z} \text{ A/m}$$

The power received by the antenna in A is:

$$P = S_t A_E = \frac{1}{2} \left| \vec{E}_t(z = 0.2 \text{ m}) \right|^2 \text{Re} \left\{ \frac{1}{\eta_2} \right\} A_E = \frac{1}{2} \frac{\left| \vec{E}_t(z = 0.2 \text{ m}) \right|^2}{|\eta_2|} \cos(\angle \eta_2) A_E \approx 3 \text{ pW}$$

Exercise 3

Given a transmitter for TV broadcasting operating at frequency $f = 18$ GHz installed on a tower with height $h = 20$ m, calculate:

- 1) The area A_1 covered by the transmitter in standard propagation conditions, i.e. assuming the refractivity gradient $dN/dh = -37$ units/km.
- 2) The refractivity gradient for which the area covered by the antenna is $A_2 = 1.5 A_1$.
- 3) The power margin (in dB) required at the transmitter to guarantee the full coverage of A_2 , assuming that the whole area is affected by a constant rain rate $R = 5$ mm/h.

Assume: the specific attenuation due to rain (dB/km) at 18 GHz, for vertical polarization and 0° link elevation angle is given by $\gamma = kR^\alpha$ where $k = 0.0771$ and $\alpha = 1.0025$.

Solution:

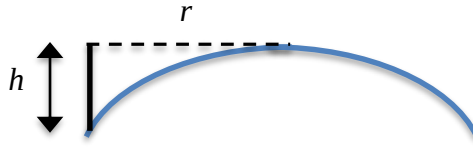
- 1) Under standard propagation conditions, the equivalent Earth radius is $R_E = 4/3 R_{earth} = 8495$ km.

The radius of the area A_1 covered by the antenna is given by the inversion of:

$$h = \frac{1}{2} \frac{r^2}{R_E} \rightarrow r = \sqrt{2hR_E} = 18.4 \text{ km}$$

Therefore the area A_1 will be:

$$A_1 = r^2 \pi = 1067.5 \text{ km}^2$$



- 2) When the propagation conditions change, the area covered by the antenna becomes:

$$A_2 = 1.5 A_1 = 1601.3 \text{ km}^2$$

which corresponds to the new radius:

$$r_2 = \sqrt{A_2/\pi} = 22.6 \text{ km}$$

The new equivalent Earth radius is given by:

$$R_E = \frac{1}{2} \frac{r_2^2}{h} = kR_{earth} = \left(\frac{1}{1 + R_{earth} \frac{dn}{dh}} \right) R_{earth} = 12743 \text{ km}$$

Inverting this equation, we obtain:

$$\frac{dn}{dh} = \left(\frac{R_{earth}}{R_E} - 1 \right) \frac{1}{R_{earth}} = -78.5 \times 10^{-6} \text{ 1/km} \rightarrow \frac{dN}{dh} = -78.5 \text{ units/km}$$

3) Under rainy conditions, the total path attenuation due to rain (assuming constant rain rate along the path) is:

$$A_R = \gamma_R r_2 = 0.387 r_2 = 8.7 \text{ dB}$$

This is also the power margin to be allocated to the transmitter to cover the whole area A_2 under rainy conditions.

Exercise 4

Consider a zenithal link (elevation angle $\theta = 90^\circ$) from a LEO satellite to a ground station, operating at $f = 19$ GHz, in which the signal goes through a uniform ice cloud (of thickness $h = 4$ km) consisting of equioriented ice needles. The specific attenuation of the ice cloud $\alpha_V = \alpha_H = \alpha = 0.025$ dB/km (V and H are associated to the vertical and horizontal wave polarization, respectively) is constant and uniform through the whole cloud and, in addition, the ice needles cause a differential phase shift (between H and V) equal to $90^\circ/\text{km}$.

Knowing that the satellite transmits a right-end circular polarization (RHCP):

- 1) Calculate the signal-to-noise ratio assuming vacuum between the transmitter and the receiver.
- 2) Determine the polarization in front of the receiver, considering the presence of the ice cloud.
- 3) Calculate the signal-to-noise ratio, considering the presence of the ice cloud.

Assumptions:

- no other sources of noise apart from the atmosphere (no cosmic background noise)
- always assume vacuum for the calculation of the wavelength
- antennas optimally pointed
- the antenna on the ground receives RHCP waves

Additional data:

- cloud temperature $T_{ice} = -5^\circ\text{C}$
- gain of the antennas (on board the satellite and on the ground): $G_T = G_R = 10$ dB
- power transmitted by the satellite: $P_T = 100$ W
- altitude of the LEO satellite: $H = 400$ km
- bandwidth of the receiver: $B = 5$ MHz
- internal noise temperature of the receiver: $T_R = 300$ K

Solution:

1) In case no clouds are present, the signal-to-noise ratio (SNR) is simply given by the link budget equation:

$$SNR = \frac{P_R}{P_N} = \frac{P_T G_T f_T \left(\lambda / 4\pi H \right)^2 G_R f_R}{kTB}$$

where k is the Boltzmann's constant (1.38×10^{-23} J/K), T is the total noise temperature (summation of T_R and the antenna noise T_A), $f_R = f_T = 1$ (antenna optimally pointed). Note that if there are no attenuating media between the transmitter and the receiver, then $T_A = 0$ K. Considering $\lambda = c/f = 0.0158$ m, and using the data available $\rightarrow SNR = 4.77 = 6.78$ dB.

2) An RHCP wave consists of two orthogonal linear components with the same amplitude and a differential phase shift of -90° . The ice cloud, overall, causes the same attenuation on both components:

$$A_{dB} = \alpha h = 0.1 \text{ dB}$$

which means that at the receiver the two components will have the same amplitude.

The cloud also causes a total differential phase shift of:

$$\Delta\phi = 90^\circ h = 360^\circ$$

Therefore at the receiver the two linear components will still produce an RHCP wave.

3) There are three effects on the SNR due to the presence of the ice cloud:

- Depolarization $\rightarrow$ no effects as the RHCP is preserved
- Additional atmospheric attenuation $\rightarrow A = 10^{-\frac{A_{dB}}{10}} = 0.977$
- Increase of the total receiver noise due to the antenna noise $T_A = T_{ice} (1-A) = 6.1 \text{ K}$

The link budget becomes:

$$SNR = \frac{P_R}{P_N} = \frac{P_T G_T f_T \left(\frac{\lambda}{4\pi H} \right)^2 G_R f_R A}{k(T_R + T_A) B}$$

We obtain:

$$SNR = 4.56 = 6.59 \text{ dB}$$