

**Radio and Optical Wave Propagation – Prof. L. Luini,
August 31st, 2021**

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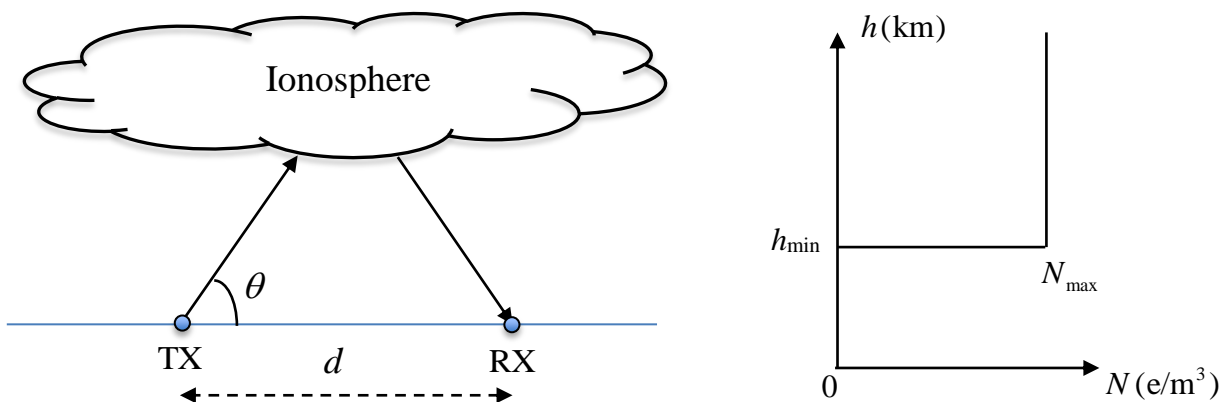
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Problem 1

Making reference to the figure below, the transmitter TX wants to reach the user RX by exploiting the ionosphere. The ionosphere can be modelled with the electron density profile sketched in the figure (right side), where $N_{\max} = 9 \times 10^{12} \text{ e/m}^3$, $h_{\min} = 100 \text{ km}$. The elevation angle is $\theta = 30^\circ$.

- 1) Calculate the power density reaching the bottom of the ionosphere, assuming that TX features an isotropic antenna with efficiency $\eta = 1$ and that the transmit power is $P_T = 100 \text{ W}$.
- 2) Calculate the distance d for TX to reach RX, assuming that the link operational frequency is $f_1 = 30 \text{ MHz}$.
- 3) Repeat the calculation at point 2) if the link operational frequency changes to $f_2 = 70 \text{ MHz}$.

Assume: the virtual reflection height $h_V = h_R$, being h_R the height at which the wave is actually reflected.



Solution:

- 1) The power density reaching the bottom of the ionosphere is:

$$S = \frac{P_T}{4\pi R^2} G_T f_T = 2 \times 10^{-10} \text{ W/m}^2$$

where both G_T and f_T are equal to 1 and $R = h_V/\sin(\theta) = 200 \text{ km}$.

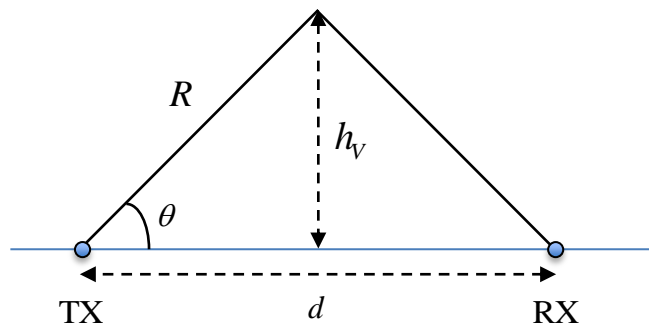
2) The elevation angle, the electron content and the link operational frequency are linked by the following relationship:

$$\cos \theta = \sqrt{1 - \left(\frac{f_c}{f}\right)^2} = \sqrt{1 - \left(\frac{9\sqrt{N_{\max}}}{f}\right)^2}$$

Solving for the frequency f , we obtain:

$$f = \sqrt{\frac{81N_{\max}}{1 - [\cos(\theta)]^2}} = 54 \text{ MHz}$$

For any frequency lower than 54 MHz, the wave will be totally reflected at the bottom of the ionosphere.



Therefore:

$$d = \frac{2h_V}{\text{tg}(\theta)} = 346.4 \text{ km}$$

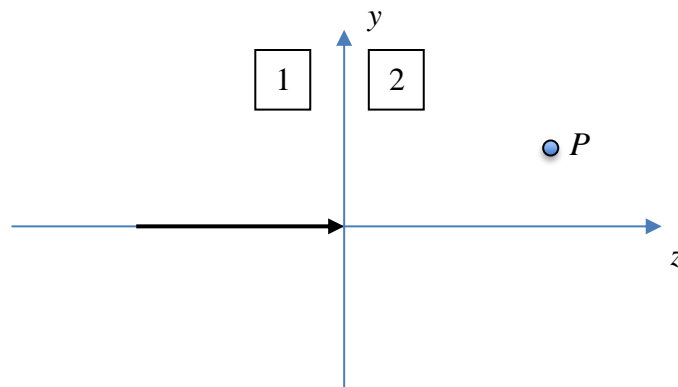
3) As $f_2 > f$, the wave crosses the ionosphere and the RX cannot be reached.

Problem 2

A plane sinusoidal EM wave (frequency $f = 10$ GHz) propagates from a perfect dielectric medium ($\epsilon_{r1} = 4$, $\mu_{r1} = 4$) into free space with orthogonal incidence. The expression for the electric field is:

$$\vec{E}(z) = \left[-\vec{\mu}_x + \frac{j}{2}\vec{\mu}_y \right] e^{-j\beta_1 z} \text{ V/m}$$

- 1) Calculate β_1 .
- 2) Determine the polarization of the incident EM wave.
- 3) Write the expression of the reflected wave.
- 4) Calculate the power received by an antenna located in point $P(x = 1 \text{ m}, y = 1 \text{ m}, z = 10 \text{ m})$ with effective area $A_E = 1 \text{ m}^2$.



Solution:

1) The phase constant in the first medium can be easily calculated as:

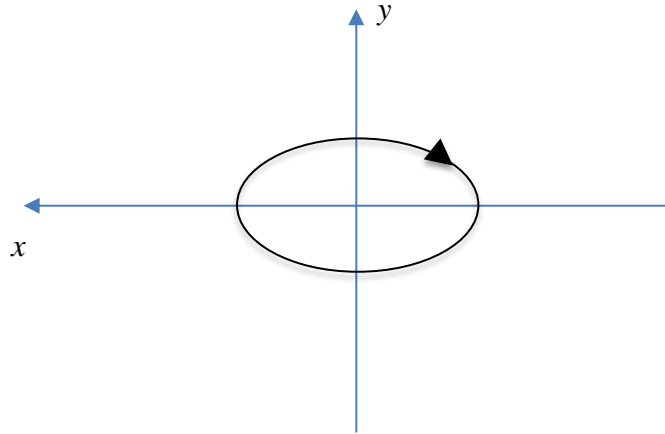
$$\beta_1 = \frac{2\pi f}{c} \sqrt{\epsilon_{r1}\mu_{r1}} = 837.8 \text{ rad/m}$$

2) The incident electric field consists of two orthogonal components, which are phase shifted and have different absolute values. As a result, the polarization of the incident wave is elliptical. To determine the rotation direction, we need to investigate the variation in time of the two components:

$$\vec{E}(0, t) = \text{Re} \left\{ \left[-\vec{\mu}_x + \frac{j}{2}\vec{\mu}_y \right] e^{-j\beta_1 z} \right\} = -\vec{\mu}_x \cos(\omega t) + \frac{1}{2}\vec{\mu}_y \cos\left(\omega t + \frac{\pi}{2}\right) \text{ V/m}$$

Thus, for $t = 0 \rightarrow \vec{E} = -\vec{\mu}_x \text{ V/m}$

Thus, for $\omega t = \pi/2 \rightarrow \vec{E} = -\frac{1}{2}\vec{\mu}_y \text{ V/m}$



The incident wave polarization is right-hand elliptical.

3) For both components, the reflection coefficient is 0, so the wave is totally transmitted into the second medium (no reflection). In fact:

$$\eta = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = 0$$

as:

$$\eta_1 = \eta_0 \sqrt{\frac{\mu_{r1}}{\epsilon_{r1}}} = 377 \, \Omega \text{ and } \eta_2 = \eta_0$$

4) As the wave is totally transmitted into medium 2, the power received at P will simply be:

$$P = \frac{1}{2} \frac{|\vec{E}|^2}{\eta_2} A_E = 0.0017 \text{ W/m}^2$$

with:

$$|\vec{E}|^2 = 1^2 + 0.5^2 = 1.25 \text{ V}^2/\text{m}^2$$

Problem 3

A transmitter for TV broadcasting operates at frequency $f = 800$ MHz, and is installed on a tower with height h . The transmit power is $P_T = 1$ W, and the antenna can be considered to be isotropic and with perfect efficiency. Calculate maximum value of h to guarantee that the minimum power reaching the receivers (positioned on the ground) at the border of the coverage area is $P_{\min} = 2 \times 10^{-11}$ W. The refractivity gradient is $dN/dh = -37$ units/km, and the receivers' antenna, whose gain is $G_R = 5$ dB, is optimally pointed to the transmitter.

Solution:

The key requirement concerns the received power, which needs to be equal to or higher than P_{\min} . The received power is given by:

$$P_R = P_T G_T f_T (\lambda / 4\pi D)^2 G_R f_R$$

assuming D to be the distance between the TX and the RX at the border of the coverage area. The link budget is further simplified by the fact that the TX antenna is isotropic and with perfect efficiency ($G_T = f_T = 1$), and that the receiver antenna is optimally pointed ($f_R = 1$). The receiver gain is $G_R = 3.16$. In order to find D , we have to impose that $P_R > P_{\min}$.

Thus:

$$P_R = P_T (\lambda / 4\pi D)^2 G_R > P_{\min}$$

Solving for D :

$$D < \frac{\lambda}{4\pi} \sqrt{\frac{P_T G_R}{P_{\min}}} = 11 \text{ km}$$

Given the propagation conditions, EM waves are bent less than the Earth's curvature, and therefore, this aspect needs to be taken into account. The equivalent Earth radius is $R_E = 4/3 R_{\text{earth}} = 8495$ km.

h can be derived as:

$$h = \frac{1}{2} \frac{D^2}{R_E} = 8.29 \text{ m}$$

For values higher than 8.29 m, the coverage area will extend too much and thus the received power will be too low.

Problem 4

Consider a zenithal link (elevation angle $\theta = 90^\circ$) from a MEO satellite to a ground station, operating at $f = 30$ GHz. The link is impaired by rain, with constant rain rate $R = 2$ mm/h (both horizontally and vertically) and height $h_R = 2$ km. The transmitted wave is circularly polarized.

1) Assuming that all rain drops are oriented horizontally (minor axis aligned with the zenithal direction), what is the wave polarization at the ground station?

For system design purpose, assume that the specific rain attenuation is given by $A_{spec} = aR^b$, with $a = 0.2291$ and $b = 0.9129$.

2) Calculate the transmit power to guarantee that, under the rainy conditions given above, the power received at the ground station is $P_R = 1$ pW. To this aim, assume that:

- antennas are optimally pointed
- the gain of the antennas on board the satellite is $G_T = 10$ dB
- the ground antenna has efficiency $\eta = 0.8$
- the ground antenna is parabolic, with diameter $D = 1.25$ m
- the altitude of the MEO satellite is $H = 8000$ km

3) Calculate the maximum bandwidth that can be used to guarantee a minimum SNR of 5 dB, given the P_R value above. To this aim, assume that:

- the internal received noise temperature is $T_R = 300$ K
- the mean radiating temperature of is $T_{rain} = 10$ °C

Solution:

1) As rain drops are oriented horizontally, their section from below will appear to be circular: any polarization will undergo the same attenuation, so the wave will not be depolarized.

2) Considering rain attenuation, the link budget is given by:

$$P_R = P_T G_T f_T (\lambda/4\pi H)^2 G_R f_R A_{rain} = P_T G_T (\lambda/4\pi H)^2 G_R A_{rain}$$

where f_T and f_R have been set to 1, as the two antennas are optimally pointed, and A_{rain} is the total rain attenuation along the path (linear scale). The latter is calculated as:

$$(A_{rain})_{dB} = aR^b h_R = 0.8627 \quad \rightarrow \quad A_{rain} = 10^{\frac{(A_{rain})_{dB}}{10}} = 0.82$$

Converting also G_T to linear scale $\rightarrow G_T = 10$. Also, the receiver antenna gain is:

$$G_R = \frac{4\pi}{\lambda^2} \eta \left(\frac{D}{2}\right)^2 \pi = 123370 \quad \rightarrow \quad (G_R)_{dB} = 50.9 \text{ dB}$$

Solving for P_T :

$$P_T = \frac{P_R}{G_T G_R (\lambda/4\pi H)^2 A_{rain}} \approx 100 \text{ W}$$

3) The total noise power in the RX is affected by the antenna noise temperature, that can be calculated as:

$$T_A = T_{rain} (1 - A_{rain}) + A_{rain} T_C = 53.3 \text{ K}$$

The SNR is:

$$SNR = \frac{P_R}{P_N} = \frac{P_R}{k(T_R + T_A)B}$$

where k is the Boltzmann's constant (1.38×10^{-23} J/K). By imposing that the minimum SNR is 5 dB (3.16), we obtain:

$$\frac{P_R}{k(T_R + T_A)B} > 3.16 \quad \rightarrow \quad B < \frac{P_R}{k(T_R + T_A)3.16} \approx 65 \text{ MHz}$$